Signal Space Representation

Procesado de Señal en Comunicaciones Inalámbricas

Curso 2023-2024



◆□▶ ◆御▶ ◆臣▶ ◆臣▶ 三臣 - のへで

Contents

Geometric Representation of Signals

Optimal Receiver in AWGN

Analysis and Simulation of Communication Signals

(ロ)、(型)、(E)、(E)、 E) の(()

Signal Space

- Internal product: $\langle s_1(t), s_2(t) \rangle = \int_a^b s_1(t) s_2^*(t) dt$
- **Orthogonality**: $s_1(t) \perp s_2(t) \Leftrightarrow \langle s_1(t), s_2(t) \rangle = 0$

Norm
$$(\sqrt{\text{Energy}})$$
: $||s(t)|| = \sqrt{\langle s(t), s(t) \rangle} = \left(\int_a^b |s(t)|^2 dt\right)^{1/2}$

• Orthonormal Basis: $\{\Psi_1(t), \Psi_2(t), \dots, \Psi_N(t)\}$ orthonormal iff

• Ortho:
$$\langle \Psi_i(t), \Psi_j(t) \rangle = 0, \qquad i \neq j$$

Normal: $\|\Psi_i(t)\| = 1, \quad \forall i$

Representation: Signal as a vector in N-dimensional space

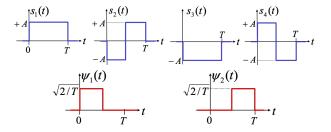
$$s(t) = \sum_{i=1}^{N} a_i \Psi_i(t)$$
 $s(t) \equiv [a_1, a_2, \dots, a_N]^T$

Orthonormal Basis

Given a set of *M* signals: {s₁(t), s₂(t),..., s_M(t)}, they can be represented by means of an orthonormal basis {Ψ₁(t), Ψ₂(t),..., Ψ_N(t)} with dimension N ≤ M

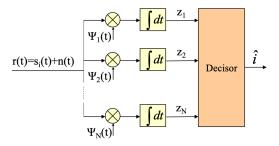
We can now analyze, study and manipulate the signals in a N-dimensional vector space

 $s_k(t) = \sum_{i=1}^N a_{k,i} \Psi_i(t) \equiv [a_{k,1}, a_{k,2}, \dots a_{k,N}]^T, \quad k = 1, \dots, M$



Optimal receiver for M symbols

- *M* equiprobable signals (symbols) in AWGN: $\{s_1(t), s_2(t), \dots, s_M(t)\}$
- Orthonormal basis $\{\Psi_1(t), \Psi_2(t), \dots, \Psi_N(t)\}$ with $N \leq M$
- Receiver based on N correlators (matched filters for $\Psi_i(t)$)
- Decisor: Minimum Distance Criterion: $\hat{i} = \arg \min_k (\|\mathbf{s}_k \mathbf{z}\|)$
 - Symbol representation: $\mathbf{s}_i = [a_{i,1}, a_{i,2}, \dots, a_{i,N}]^T$
 - Noise representation $\mathbf{n} = [n_1, n_2, \dots, n_N]^T$
 - Received Signal representation: $\mathbf{z} = [z_1, z_2, \dots, z_N]^T = \mathbf{s}_i + \mathbf{n}$

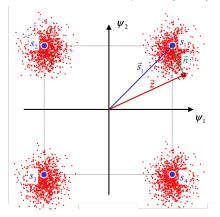


▲□▶ ▲□▶ ▲□▶ ▲□▶ □ のへで

Receiver Analysis

• Received signal: $\mathbf{z} = \mathbf{s}_i + \mathbf{n}$

- **• n** is a Gaussian random vector $\mathcal{N}(\mathbf{0}, \sigma^2 \mathbf{I})$
- Independent components with zero mean and variance $\sigma^2 = N_0/2$
- ▶ If the symbol s_i is transmitted: $\mathbf{z}|\mathbf{s}_i \sim \mathcal{N}(\mathbf{s}_i, \sigma^2 \mathbf{I})$



Optimal Decoding

- Given the received signal (z = s_i + n), we need to determine which symbol was transmitted
- This is a classification problem:
 - Minimum error criterion (MAP): $P(\mathbf{s}_i | \mathbf{z}) = \frac{f(\mathbf{z} | \mathbf{s}_i) P(\mathbf{s}_i)}{f(\mathbf{z})}$
 - For equiprobable symbols reduces to maximizing $f(\mathbf{z}|\mathbf{s}_i)$ (ML)
 - In the AWGN case, it further reduces to the minimum distance criterion: î = arg min_k (||s_k z||)

- The probability of error will depend on:
 - Distance among symbols: $d_{i,j} = \|\mathbf{s}_i \mathbf{s}_j\|$
 - Noise variance: $\sigma^2 = N_0/2$
 - Mean Bit Energy: $E_b = E_s / \log_2(M)$
 - Symbol Energy: $E_{s_i} = ||\mathbf{s}_i||^2$
 - Finally: $P_e = g(d, \sigma) = h(E_b, N_0)$

Analysis and Simulation of Communication Signals

Complete System:

- Includes Modulation, Amplifiers, Filters, Channel, Demodulation . . .
- Sampling Frequency: $f_s \ge 2f_{max}$: Impractical in most of the cases

Low-Pass Equivalent:

- Analysis of the baseband signals
- The effect of channel and TX/RX parts is modeled by their low-pass equivalent
- Sampling Frequency moderate: In practice $f_s = R_s L$ with L > 2
- Equivalent Discrete System:
 - Analysis in Signal Space
 - Sampling Frequency: $f_s = R_s$ One sample per symbol

Consider WiFi: $f_c = 2.4 GHz$, W = 20 MHz

▲□▶ ▲□▶ ▲□▶ ▲□▶ ■ ● ●